

# Introduction & Motivation

EE564: Digital Communication and Coding Systems

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**USC** University of  
Southern California

# Course Topic (from Syllabus)

- **Overview of Comm/Coding**
- Signal representation and Random Processes
- Optimal demodulation and decoding
- Uncoded modulations, demod, performance
- Classical FEC
- Modern FEC
- Non-AWGN channels (intersymbol interference)
- Practical consideration (PAPR, synchronization, spectral masks, etc.)

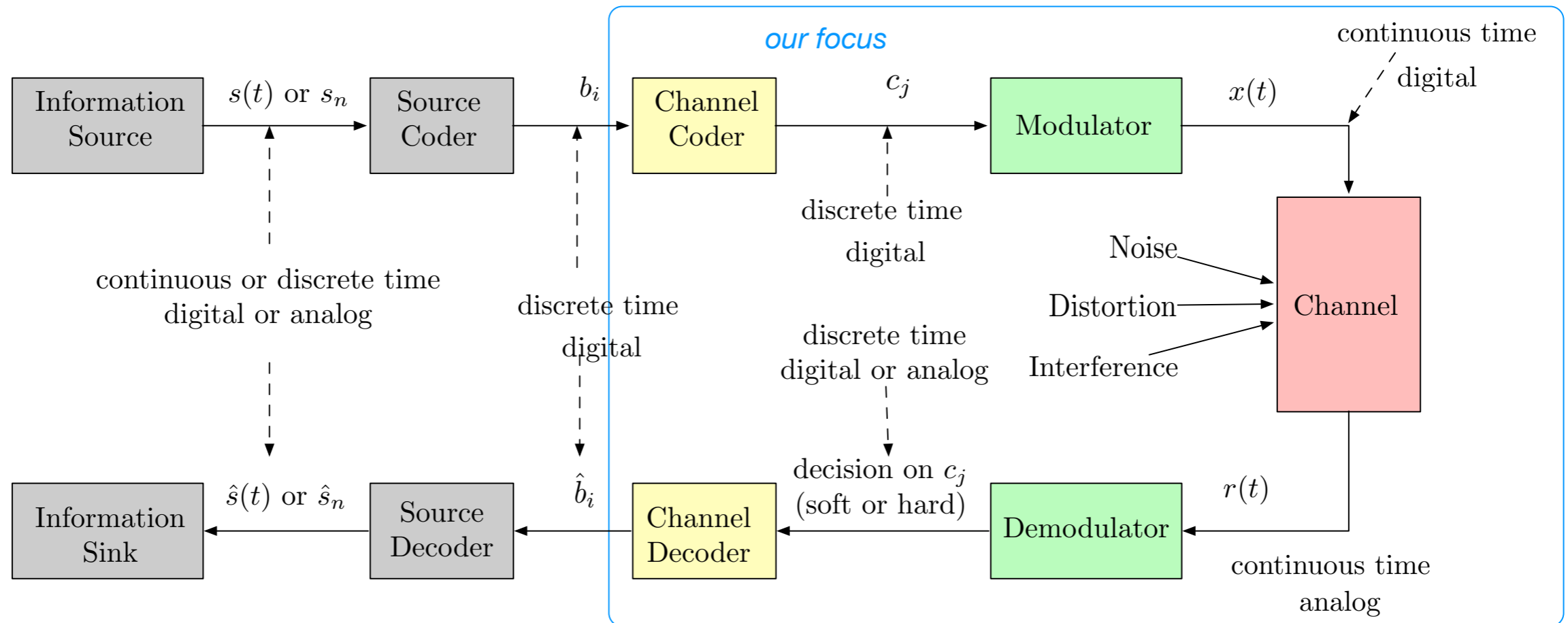
# Overview Topics

- Why Digital Comm? Why not analog?
- The digital comm system block diagram
  - Source model and entropy
  - Separation and channel capacity (mutual information)
  - Modulations, Channels, Soft vs. Hard Decision Information
- Performance measures
- Overview of Coding
- More Channels

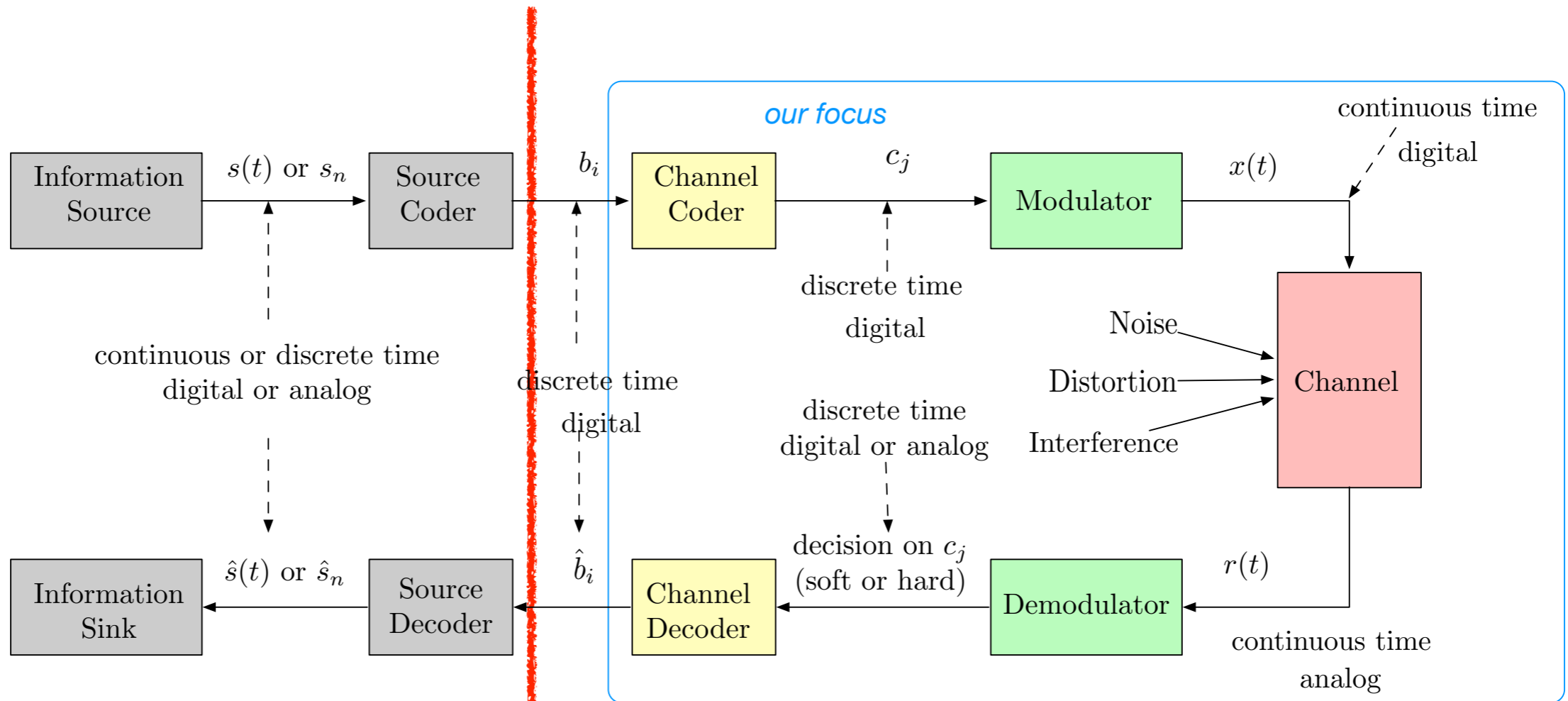
# Digital vs. Analog Communications

- Digital comm = send one of a finite number of signals at the transmitter (most modern systems are digital)
- Analog comm = send messages on the continuum (e.g., analog FM radio)
- Why Digital?
  - Exploits digital processing resources (Moore's Law) via ASICs, FPGA, DSP, etc.
  - More robustness and better fidelity — via use of memory in encoding/decoding
  - Control the amount of degradation from source to sink
  - Security (encryption)
  - Easier to share resources: multiplexing, routing, multiple access, multimedia
  - Advantages in multi-hop systems — alleviates distortion accumulating over hops

# Digital Comm. Block Diagram



# Separation Theorem

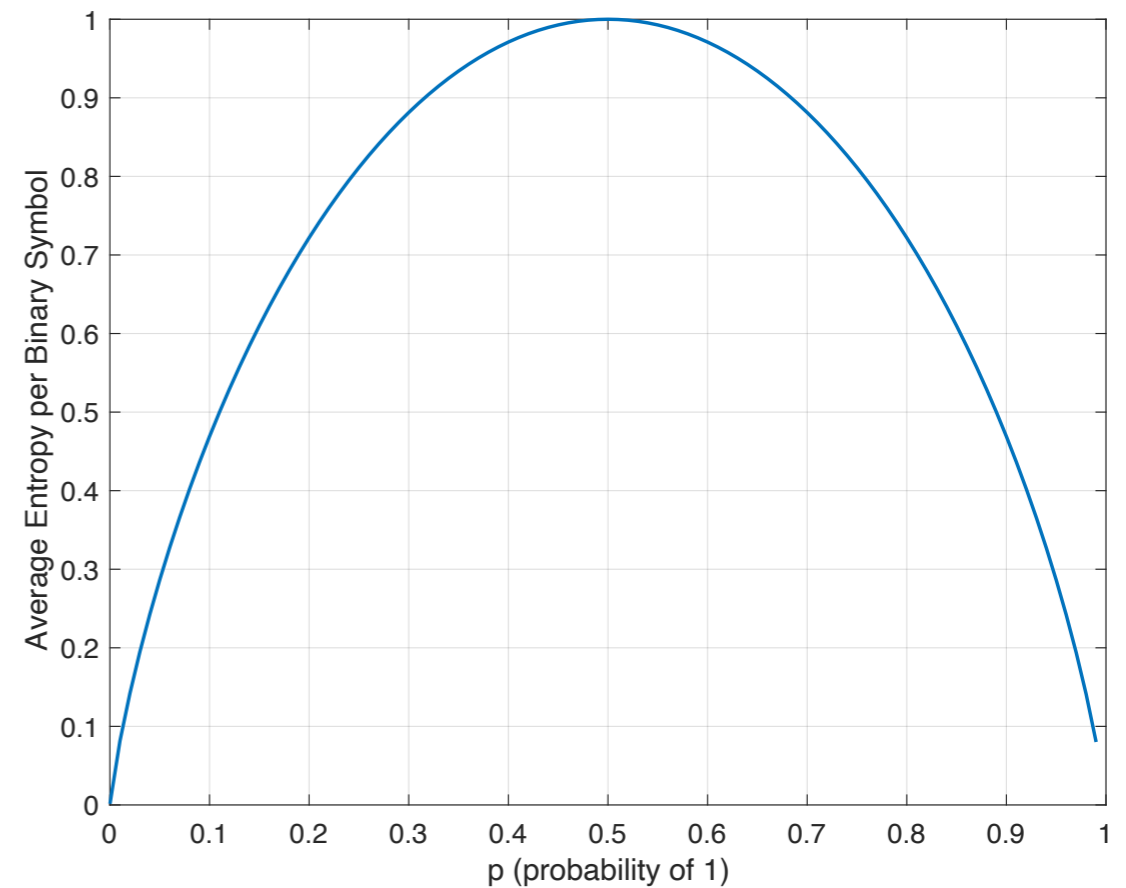
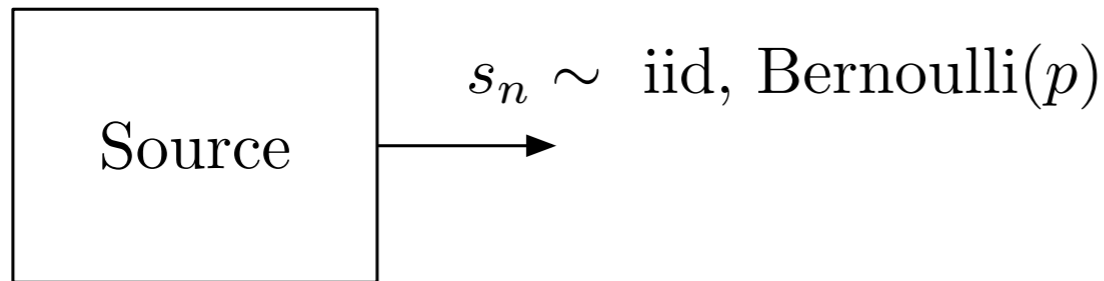


*Sample, quantize, compress this to an acceptable distortion with minimal info rate  $R$*

*Get bits through this at an info rate  $R < C$ ,  
 $C =$  channel capacity*

*There is no benefit to combining these tasks if the encoding length for each encoder can be arbitrarily large*

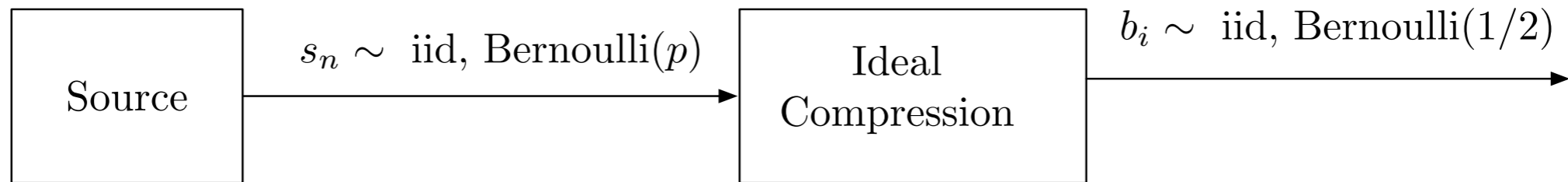
# Source Model: Binary Memoryless



## Entropy of the source

$$H(s_n(u)) = p \log_2 \left( \frac{1}{p} \right) + (1 - p) \log_2 \left( \frac{1}{1 - p} \right) \quad (\text{bits/source symbol}).$$

# Lossless Compression



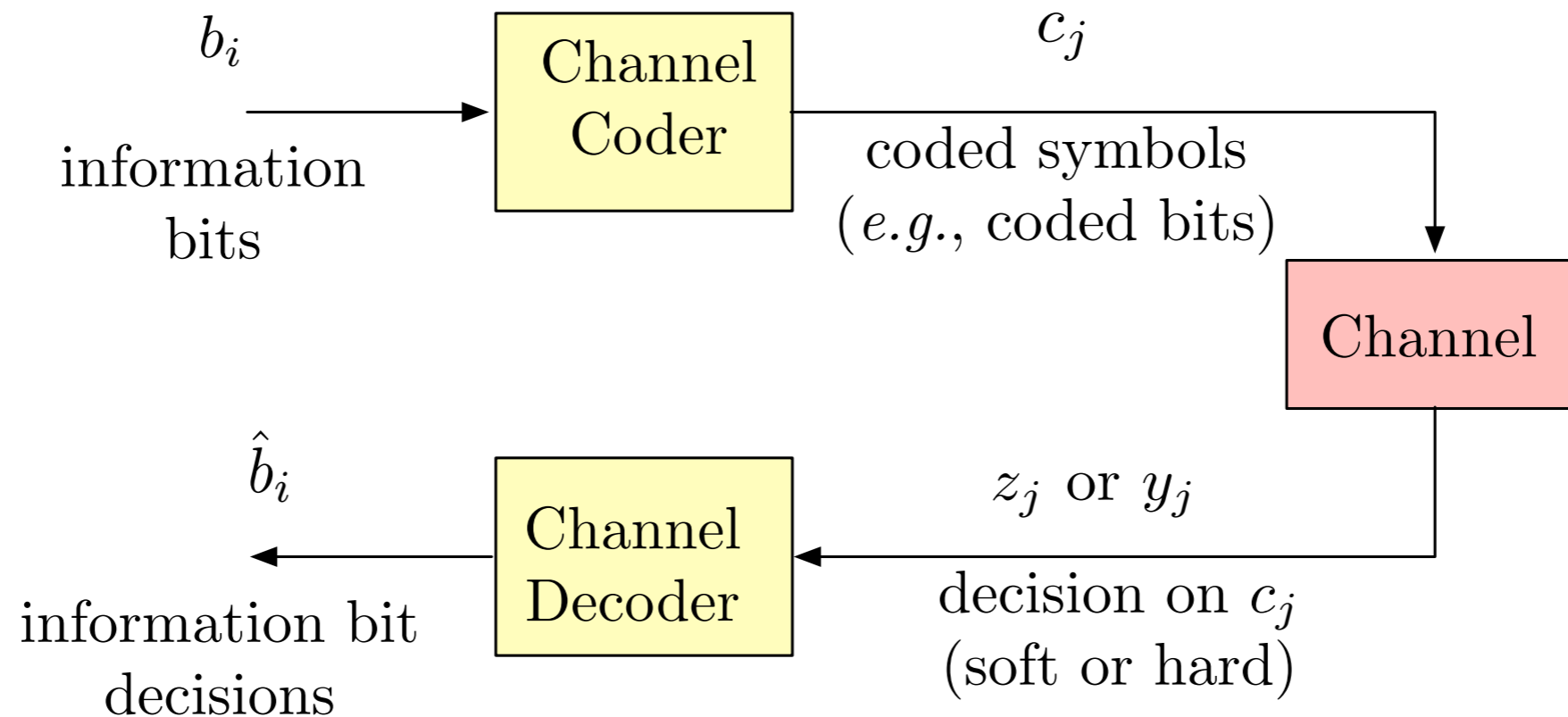
- (Lossless) Source coding theorem:
  - “Source can be compressed to its Entropy and no further”
    - For asymptotically large encoding block size
    - H values of b for each value of s

For EE564, the effective information source is b and it is iid, Bernoulli(0.5)

*(we will consider sources with  $p \neq 0.5$  for iterative decoding)*



# Coding Block Diagram



This simplified model abstracts the modulation-demodulation and details of the waveform channel

*Simplified model is used to study coding*

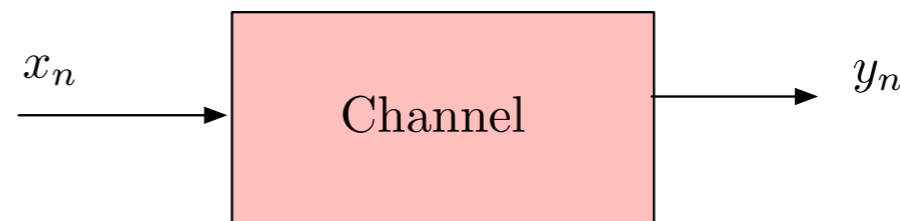
# Channel Capacity

## Mutual Information

$$\begin{aligned} I(x(u); y(u)) &= \sum_y \sum_x p_{x(u), y(u)}(x, y) \left[ \log_2 \left( \frac{p_{x(u), y(u)}(x, y)}{p_{x(u)}(x) p_{y(u)}(y)} \right) \right] \\ &= \sum_y \sum_x p_{x(u), y(u)}(x, y) \left[ \log_2 \left( \frac{1}{p_{x(u)}(x)} \right) - \log_2 \left( \frac{1}{p_{x(u)|y(u)}(x|y)} \right) \right] \\ &= \underbrace{H(x(u))}_{\text{Entropy in } x(u)} - \underbrace{H(x(u)|y(u))}_{\text{Entropy in } x(u) \text{ given } y(u)} \end{aligned}$$

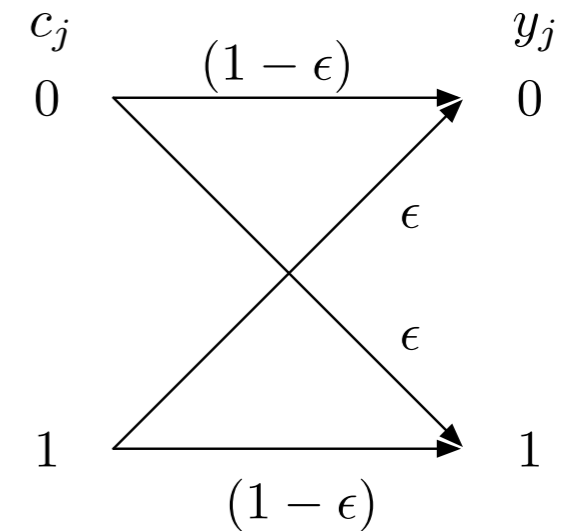
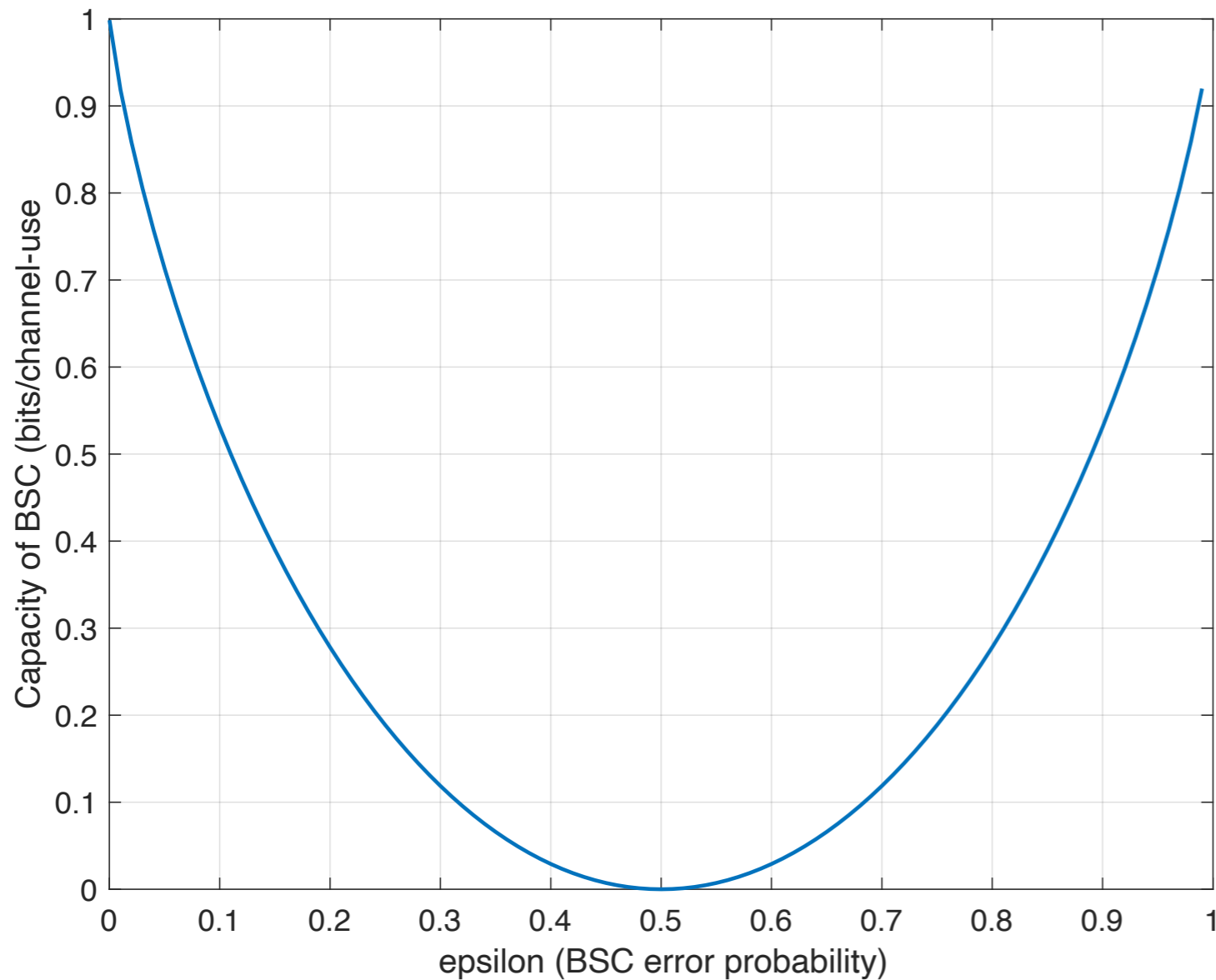
## Channel Capacity for Memoryless Channel

$$\max_{p_{x(u)}(\cdot)} I(x(u); y(u))$$



$$P(\mathbf{y}|\mathbf{x}) = \prod_n P(y_n|x_n)$$

# Channel Capacity: Binary Symmetric Channel

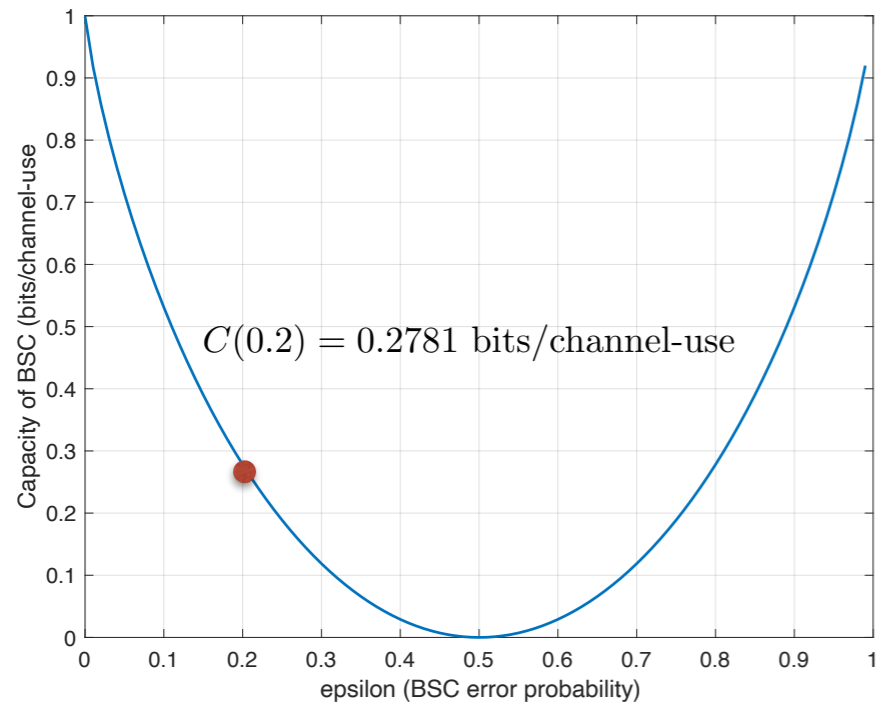


labels:  $p_{y_j(u)|c_j(u)}(y_j|c_j)$

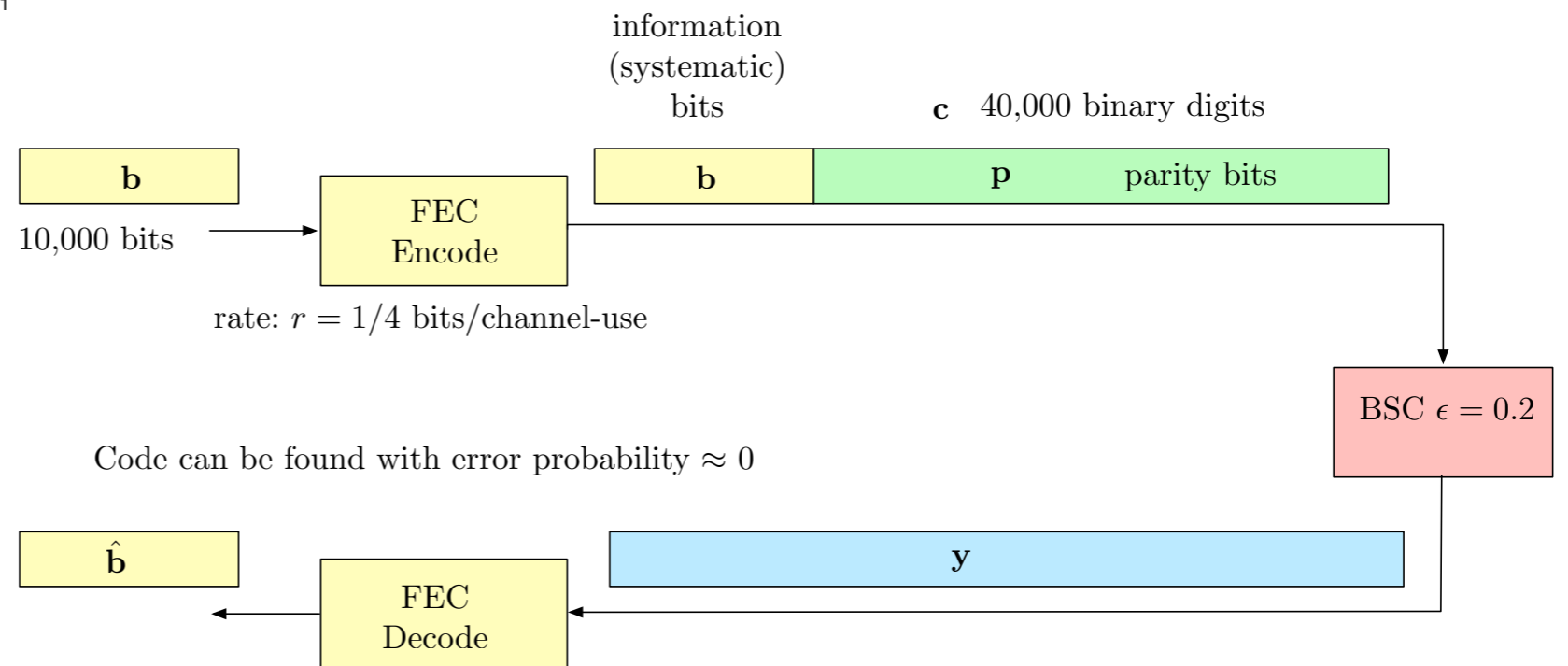
*BSC is a special case of discrete memoryless channel (DMC)*

$$C(\epsilon) = 1 - H(\epsilon) = 1 + \epsilon \log_2 \left( \frac{1}{\epsilon} \right) + (1 - \epsilon) \log_2 \left( \frac{1}{(1 - \epsilon)} \right)$$

# Interpreting BSC Capacity

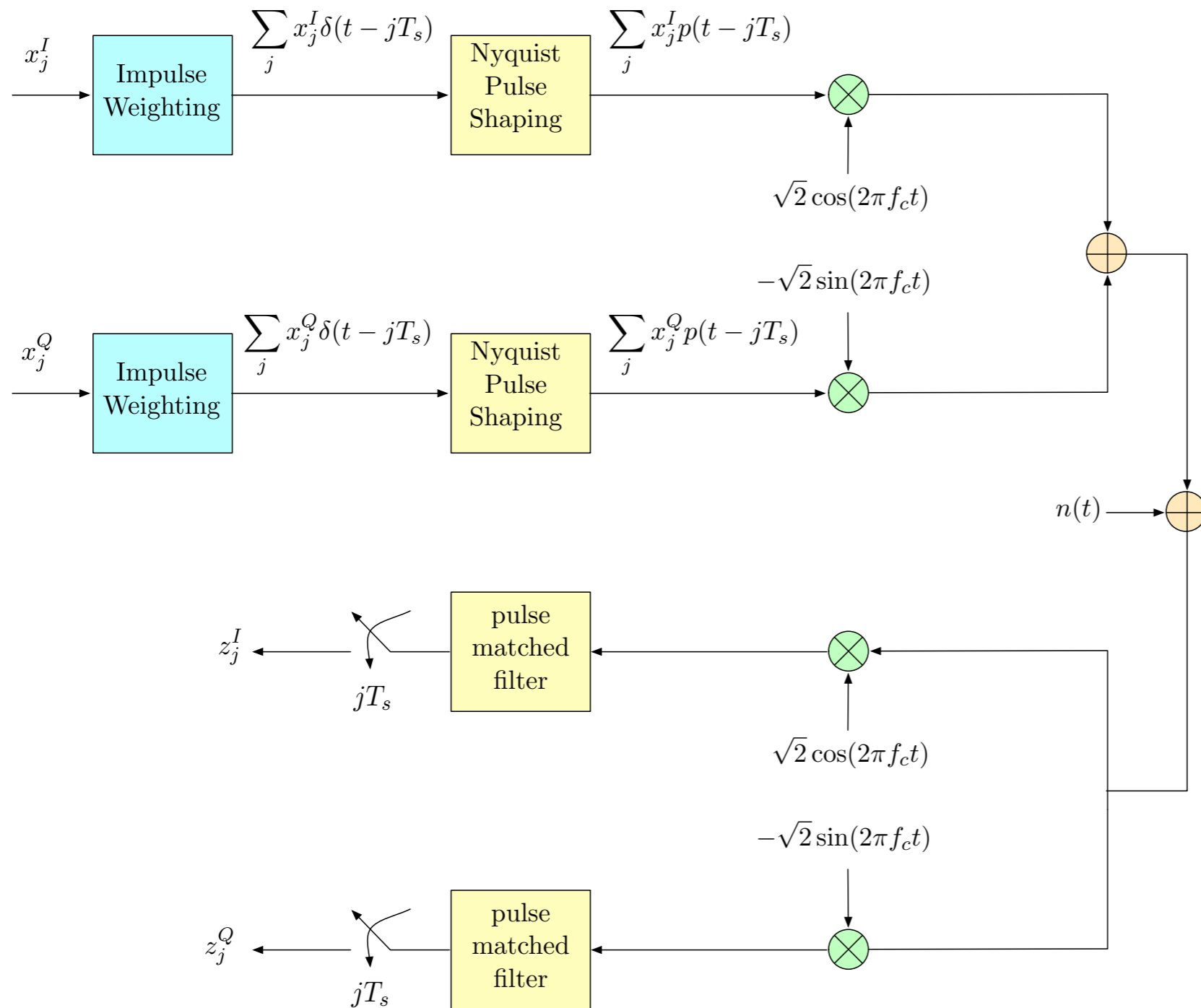


*Example of what capacity means*



*capacity is achieved (approached) through coding*

# Typical In-Phase/Quadrature Digital Modulation

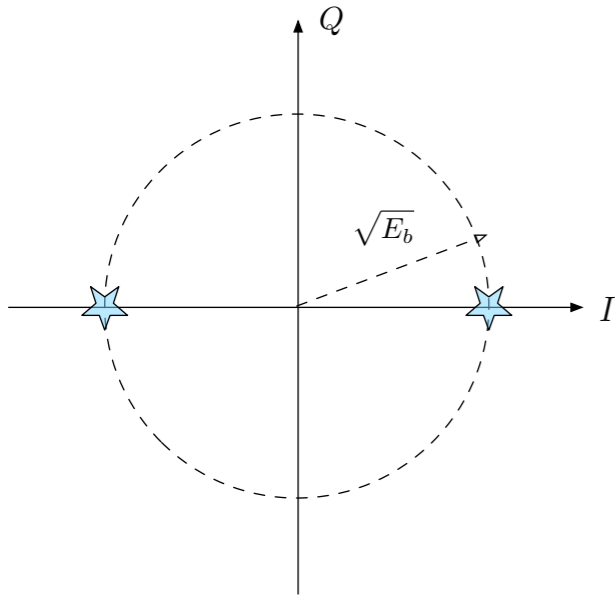


*This is approach converts a waveform channel to a 2-dimensional channel vector*

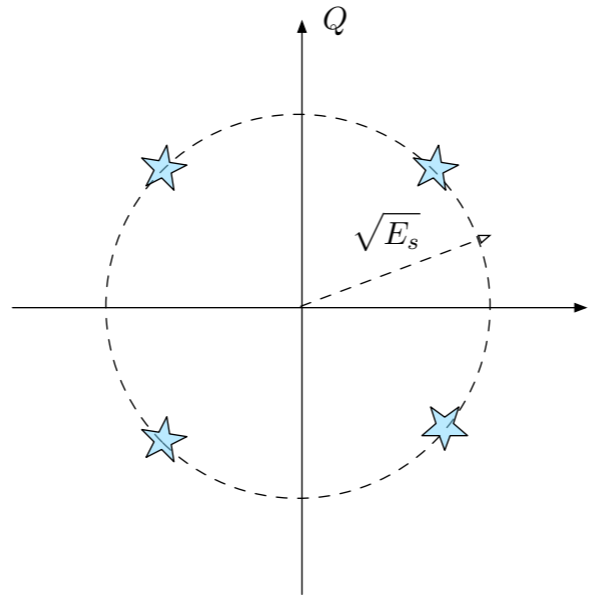
$$R_p(t) = p(t) * p(-t)$$

$$R_p(jT_s) = \delta_K(j)$$

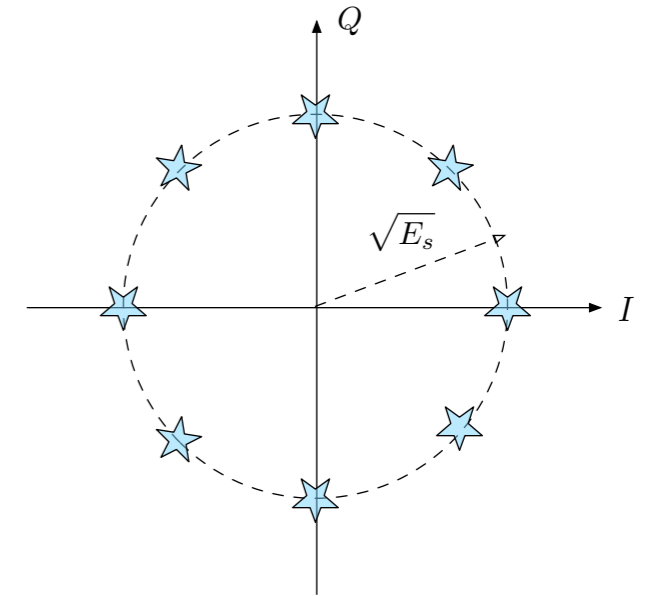
# Common I/Q Digital Modulations



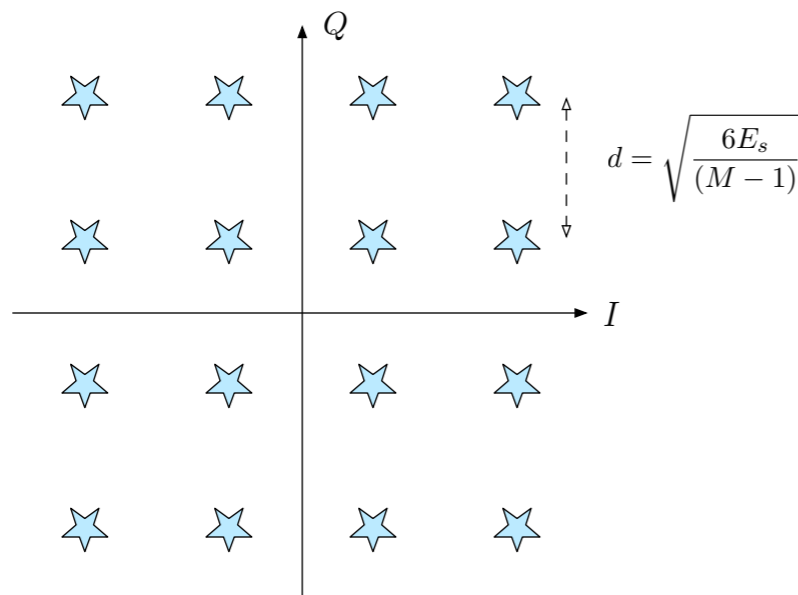
Binary Phase Shift Keying  
(BPSK)



Quadrature Phase Shift Keying  
(QPSK)



8-ary PSK (8-PSK)

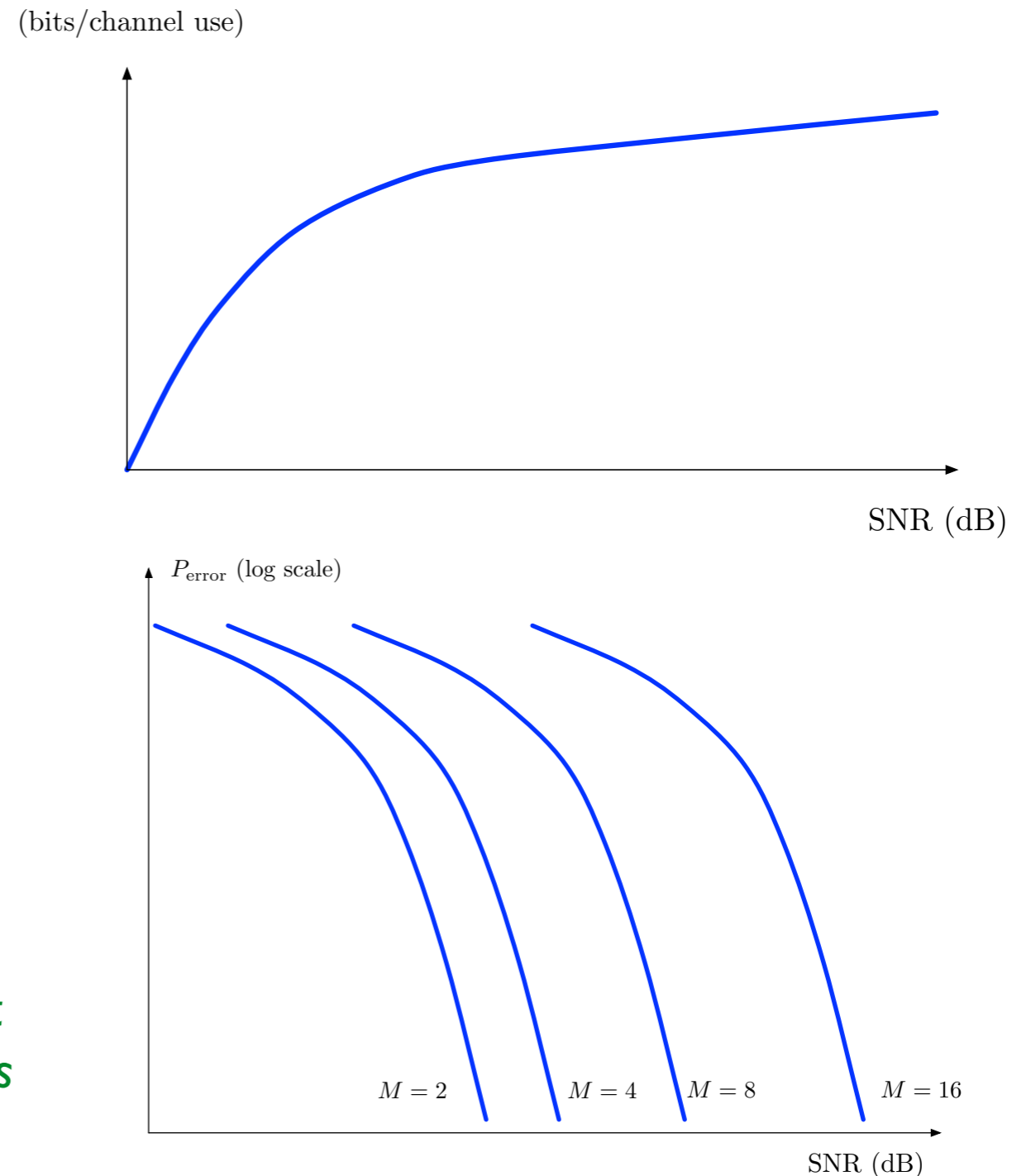


16 Quadrature Amplitude Modulation  
(16QAM)

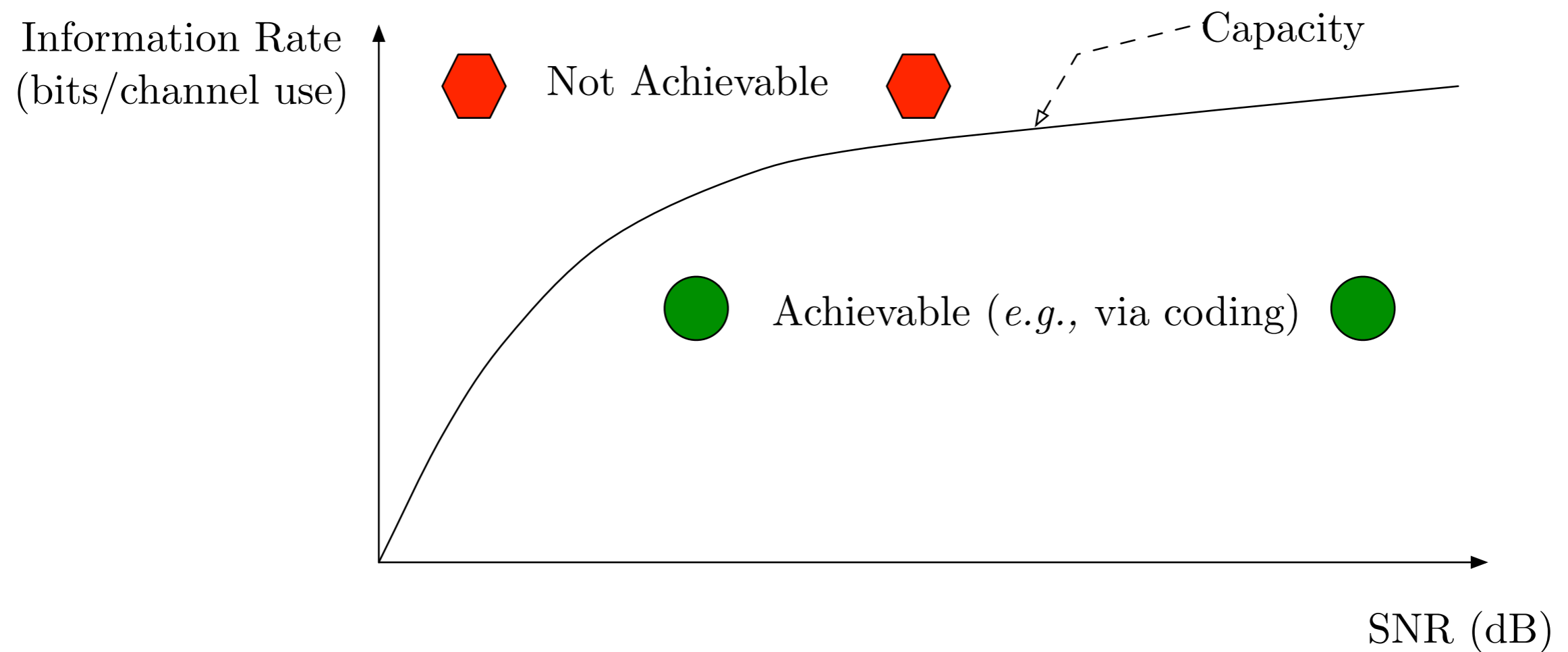
# Trade-offs with M?

- Dimension = Time \* Bandwidth
- The I/Q constellations use 2 dimensions
- As you increase M
  - More bits/channel use ( $\log_2(M)$ )
  - Points get closer for fixed energy

*These are the two types of performance plots that we use to evaluate coding and modulation schemes*



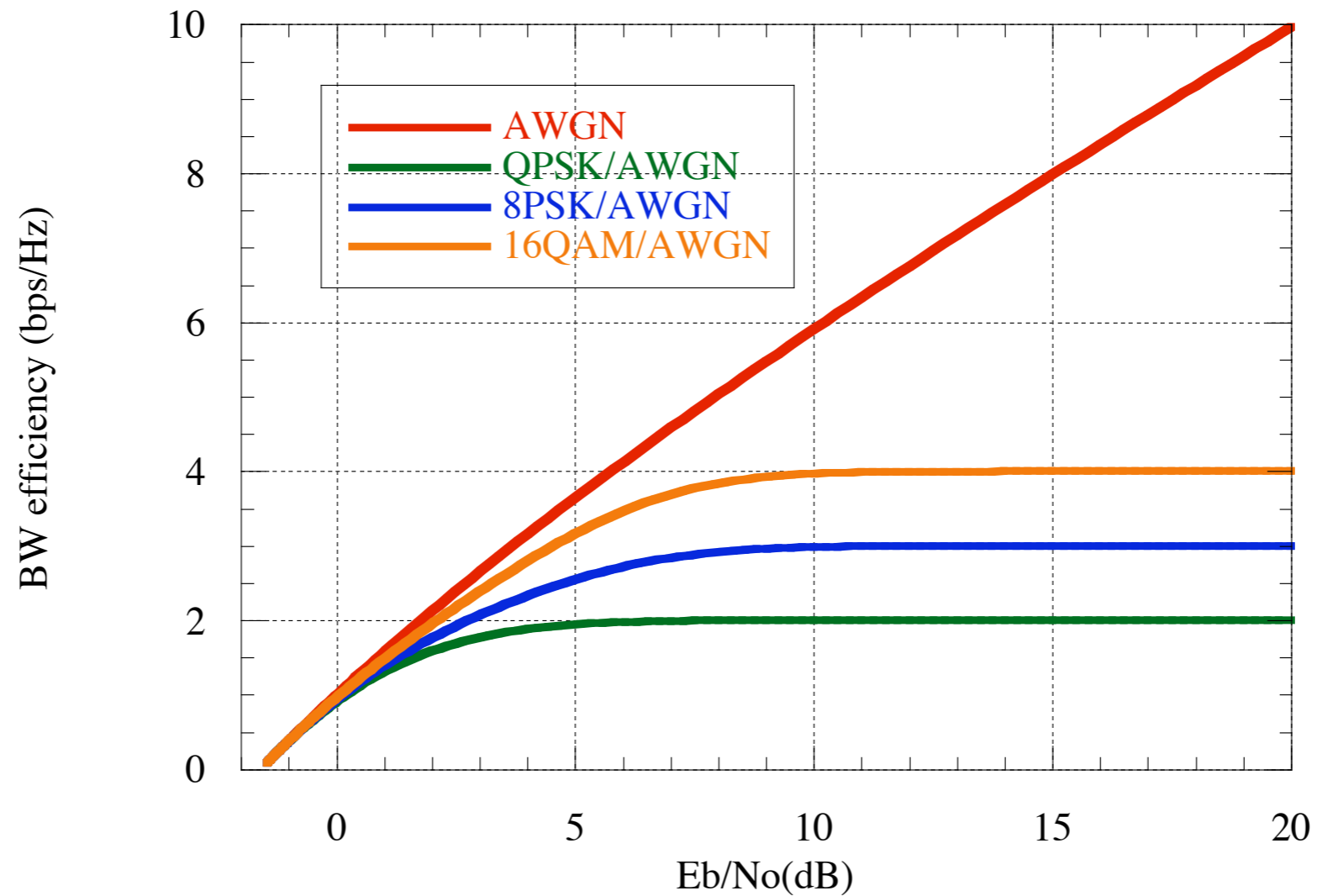
# Throughput vs. SNR Trade (IT)



*Channel capacity shows up on this type of plot as regions of achievable performance*

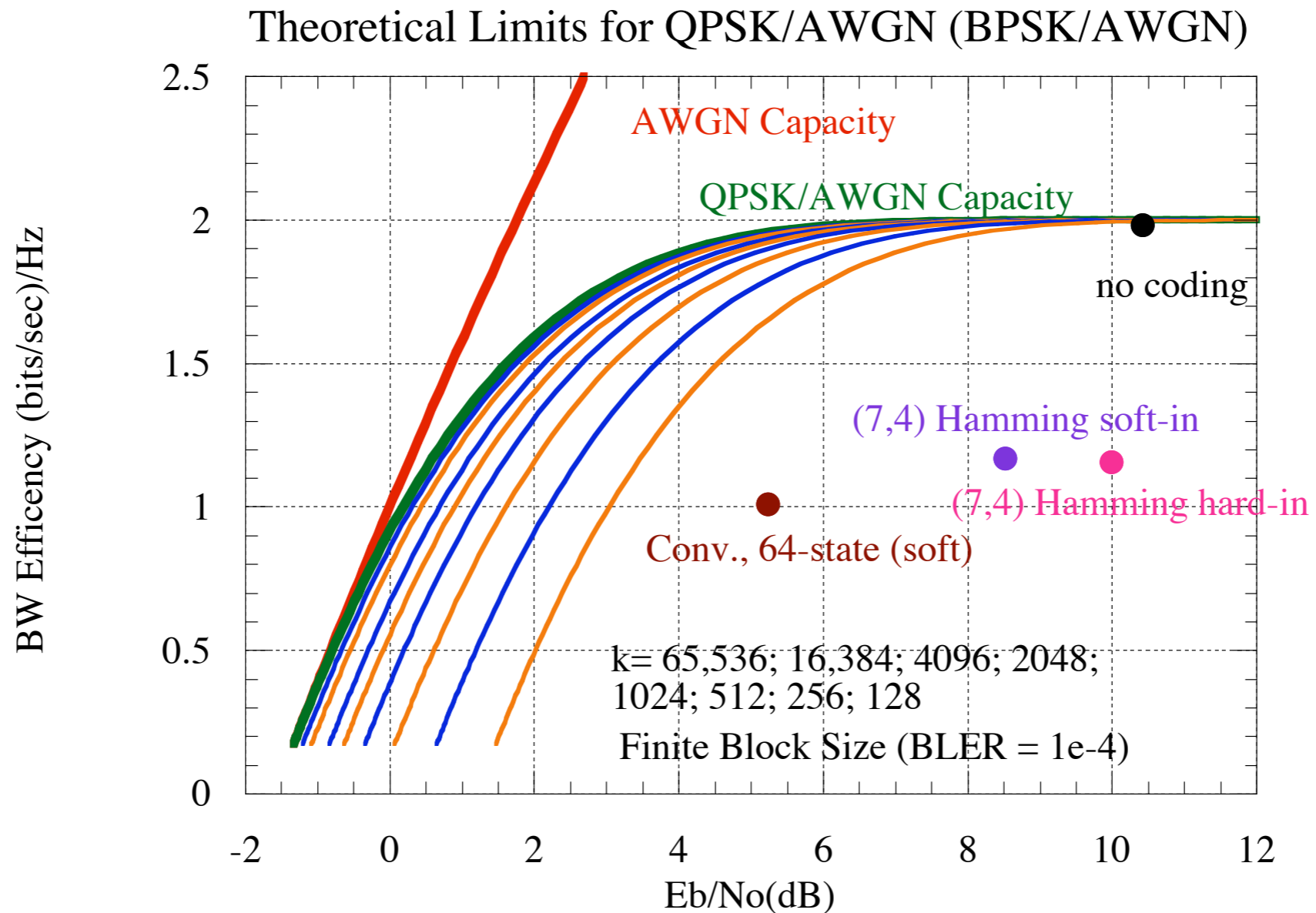


# Throughput vs. SNR Trade (IT)



*Example channel capacity curves for AWGN channels with and without modulation constraints*

# Throughput vs. SNR Trade



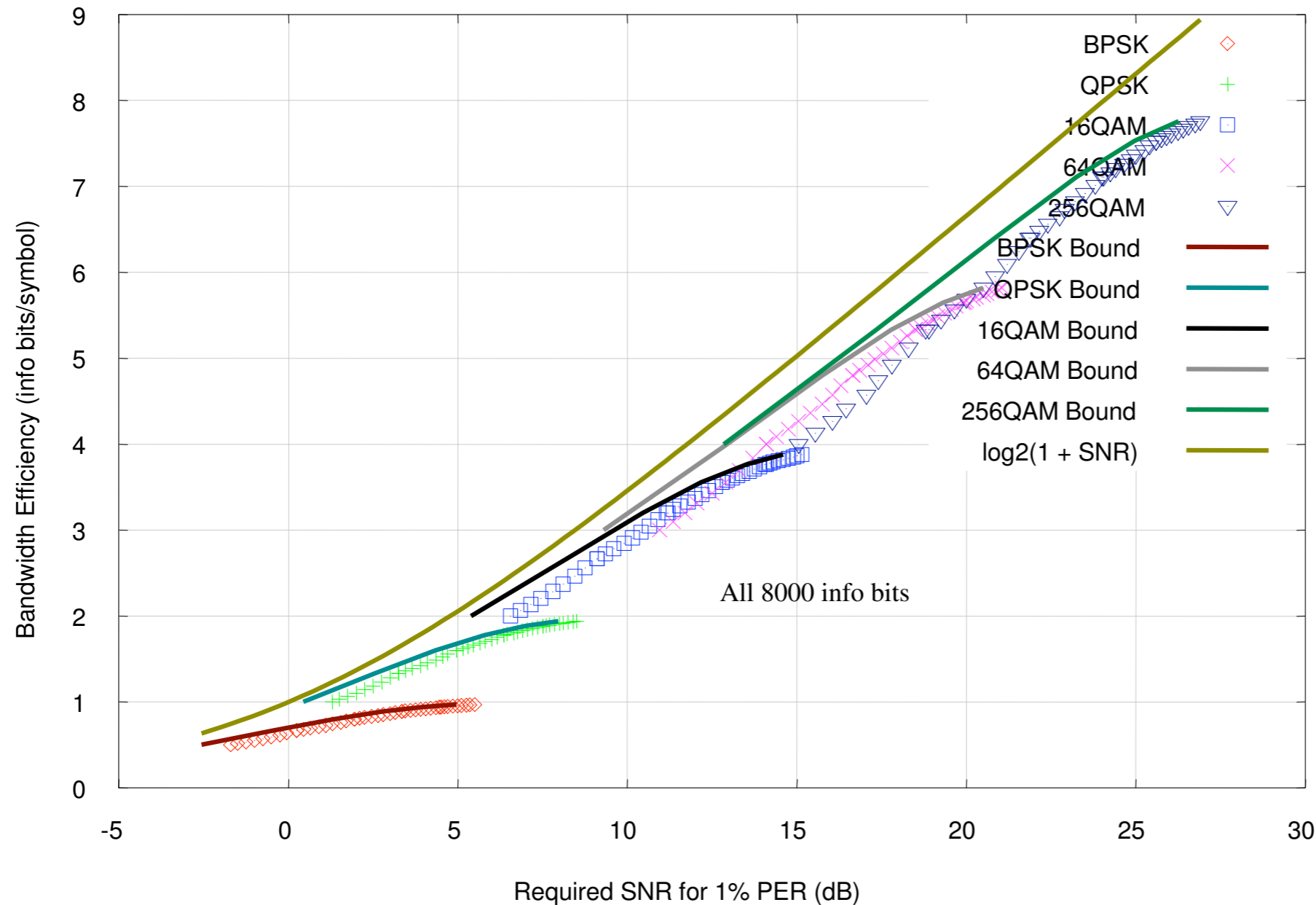
*How some common coding schemes with BPSK modulation compare to capacity*

# Throughput vs. SNR Trade

September 2004

doc.: IEEE 802.11-04/0953r4

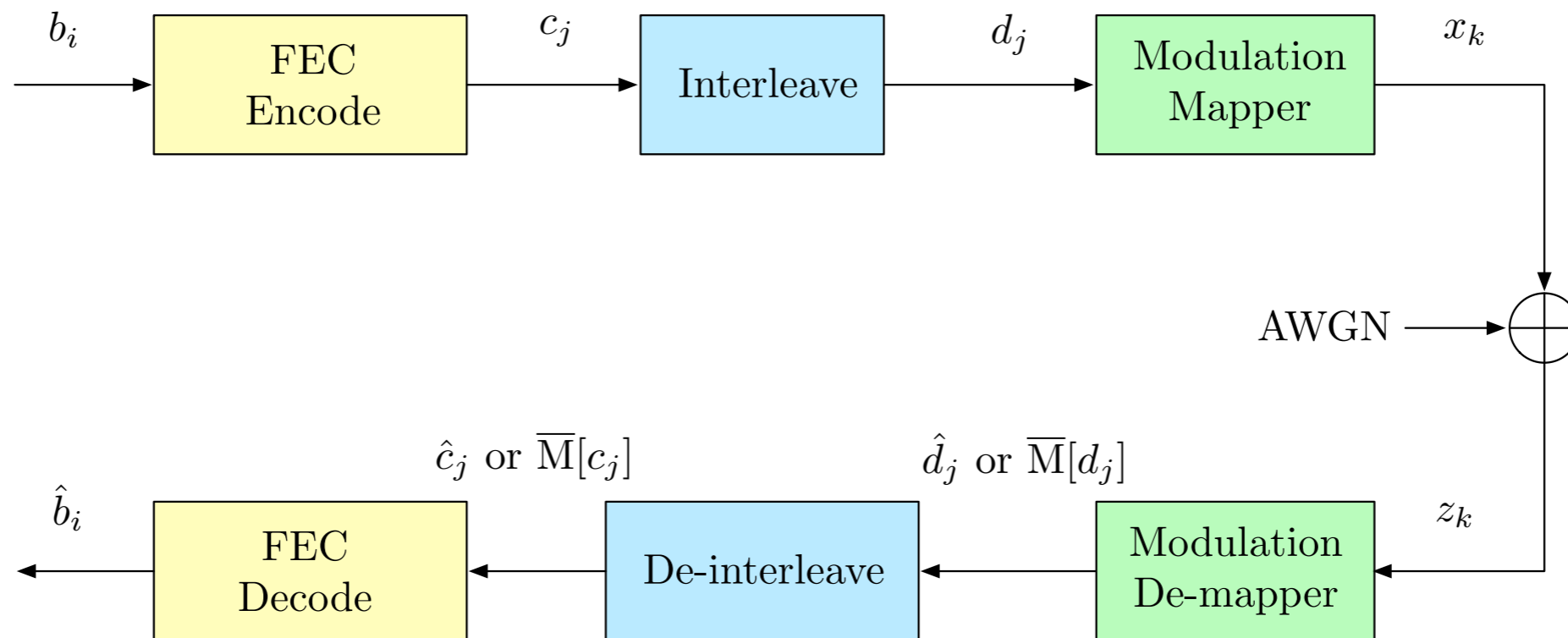
## AWGN Perf.: Comparison with Bound



*An example of a standards contribution based on these same concepts*



# Bit-Interleaved Coded Modulation (BICM)

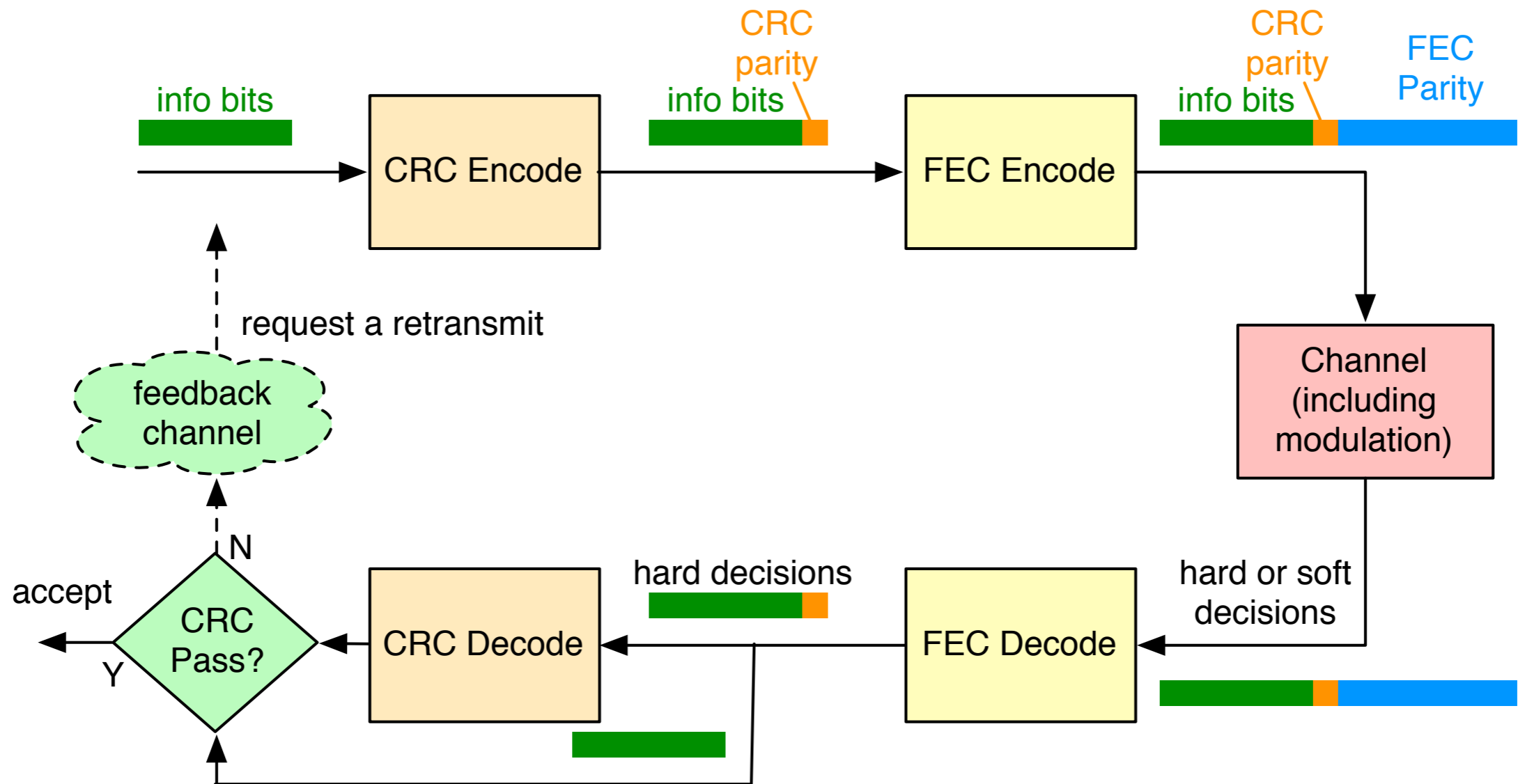


Most common approach to coding & modulation used in practice  
(you will simulate a BICM system with a modern code this semester)

# Summary of Channel Coding

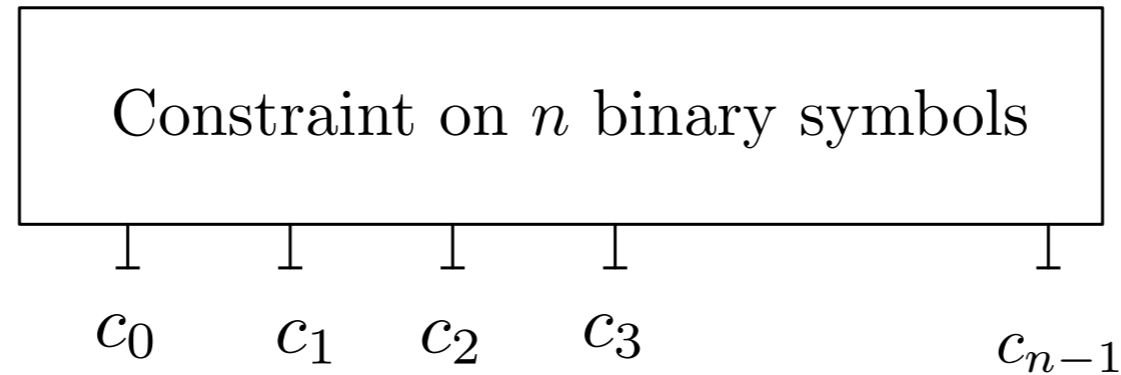
- **Forward Error Control/Correction Coding (FEC)**
  - As described previously — add redundancy and send across channel
- Error Detection Coding (aka CRC = Cyclic Redundancy Check)
  - Detect if an error has occurred on the channel, but no correction
- Automatic Repeat Request (ARQ)
  - “Hey, I did not get that, send it again!”

# Typical Use of Coding in Modern System



Hybrid ARQ (H-ARQ) System

# Codes as Constraints on Variables



Only  $2^k$  of the  $2^n$  ( $n \times 1$ ) binary vectors are in the code

- **Repetition Code (equality constraint)**
  - *all  $n$  bits are the same*
- **Single Parity Check Code (SPC code)**
  - *only patterns with even number of 1s*

# Example: Repetition Code

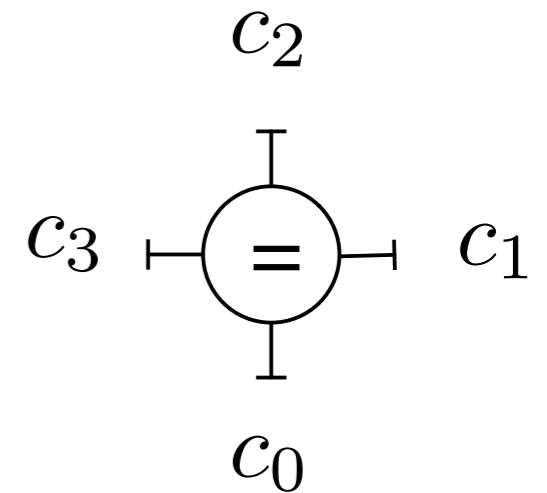
Codewords for  $n = 4$ : 0000 1111

Number of codewords = 2, so  $k = 1$

rate =  $1/n$  (info bits per channel use)

Encoder:

take one information bit in and output  $n$  copies of this bit

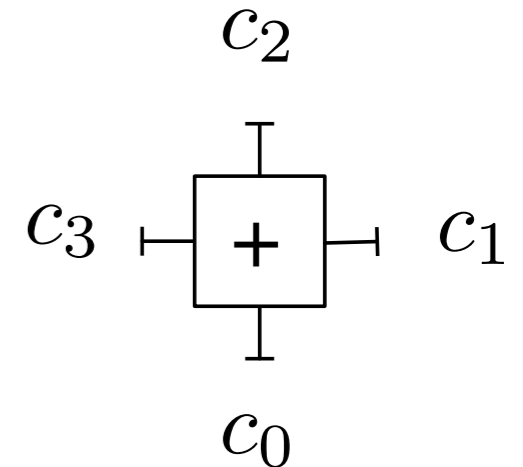




# Example: Single Parity Check Code

Codewords for  $n = 4$ :

0000	0101
0011	1001
1100	0110
1010	1111



Number of codewords = 8 , so  $k = 3 = n-1$

rate =  $(n-1)/n$  (info bits per channel use)

**Encoder:**

take  $n-1$  information bit in and these plus one parity bit  
which is the mod 2 sum of the input bits

# Example: (7,4) Hamming Code

$$\mathbf{H} = \begin{bmatrix} 1 & 1 & 0 & 1 & 1 & 0 & 0 \\ 1 & 0 & 1 & 1 & 0 & 1 & 0 \\ 0 & 1 & 1 & 1 & 0 & 0 & 1 \end{bmatrix}$$

$$\mathbf{H}\mathbf{c} = \mathbf{0}$$

$c_0$     $c_1$     $c_2$     $c_3$     $c_4$     $c_5$     $c_6$

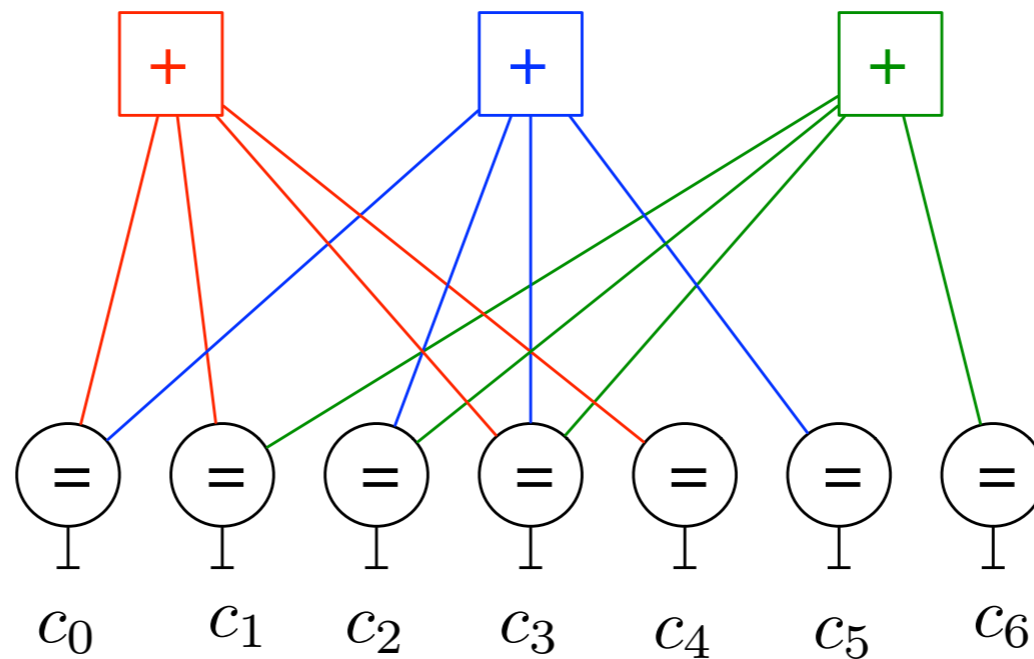
Linear Block Code (“Multiple Parity Check Code”)

*All three SPCs must be satisfied simultaneously*

# Example: (7,4) Hamming Code

$$\mathbf{H} = \begin{bmatrix} 1 & 1 & 0 & 1 & 1 & 0 & 0 \\ 1 & 0 & 1 & 1 & 0 & 1 & 0 \\ 0 & 1 & 1 & 1 & 0 & 0 & 1 \end{bmatrix}$$

*Parity Check Graph  
or Tanner Graph*



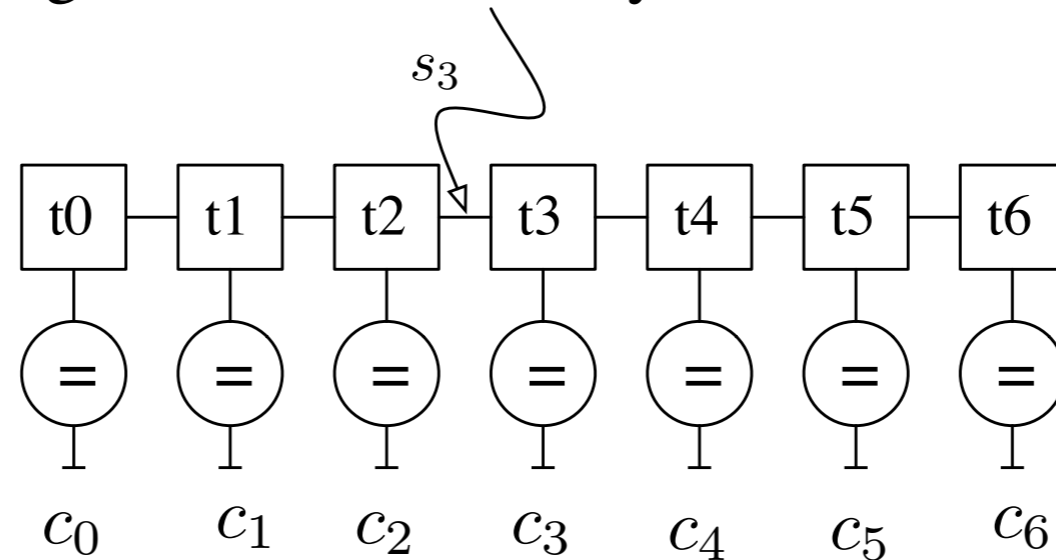
*All local constraints must be satisfied simultaneously*

# Example: (7,4) Hamming Code

$$\mathbf{H} = \begin{bmatrix} 1 & 1 & 0 & 1 & 1 & 0 & 0 \\ 1 & 0 & 1 & 1 & 0 & 1 & 0 \\ 0 & 1 & 1 & 1 & 0 & 0 & 1 \end{bmatrix}$$

e.g., hidden, non-binary state variable

*Parity Check Trellis  
Graphical Model*



*All local constraints must be satisfied simultaneously*

# Example: Low Density Parity Check (LDPC) Code

*Just a very large (multiple) parity check code with mostly 0s*

$$\mathbf{H} = \begin{bmatrix} 1 & 0 & \dots & 1 & 0 & 0 \\ 0 & 0 & \dots & 0 & 1 & 0 \\ \vdots & \vdots & \ddots & \vdots & \vdots & \\ 1 & 1 & \dots & 0 & 0 & 0 \end{bmatrix}$$

*number of 1s = number of bits in first SPC*

*number of 1s = number of SPCs second code bit is involved in*

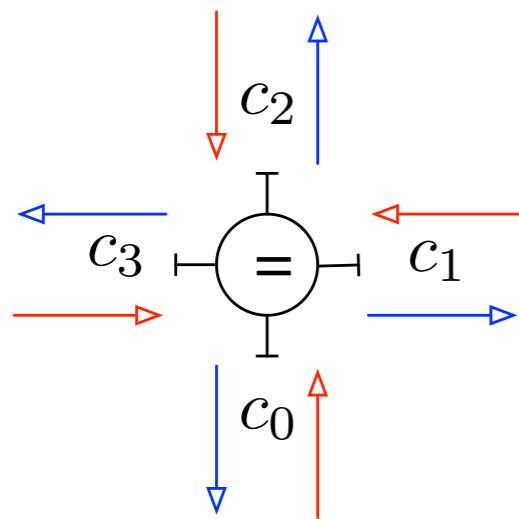
*A systematic way to build codes with very large block size*

# Example: Low Density Parity Check (LDPC) Code

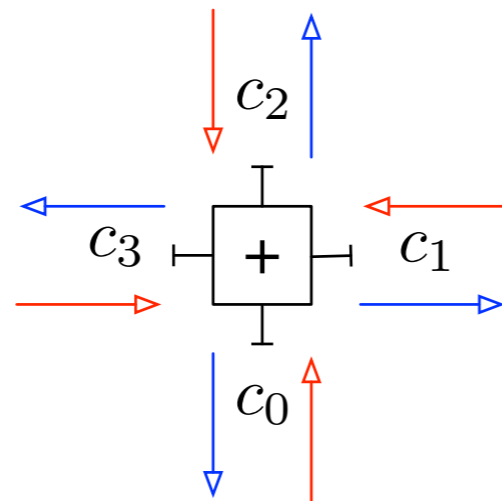
*The trick is in the decoding algorithm:*

*Repeatedly do “soft-in/soft-out (SISO)” decoding of each local code and exchange these soft decisions (messages, beliefs, metrics)*

*ITERATE until things look good!*



Equality Constraint SISO



SPC SISO

*SISO rule  
(message update rule)  
depends on code constraint*

- Incoming soft-decision information
- Outgoing soft-decision information

# Example: Low Density Parity Check (LDPC) Code

*This simple construction with this decoding approach can approach channel capacity with large block sizes*

# Summary of Modern FEC

- **Construct large codes (big n) by connecting simple, local (or constituent) codes via pseudo-random permutations**
- **Iteratively decoding**
  - ***Run SISO decoding for each local code***
  - ***Exchange soft-information between local code SISOs***

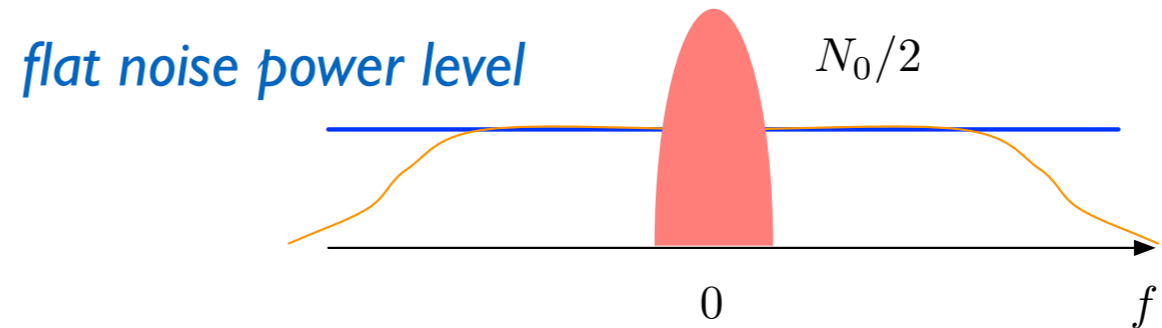
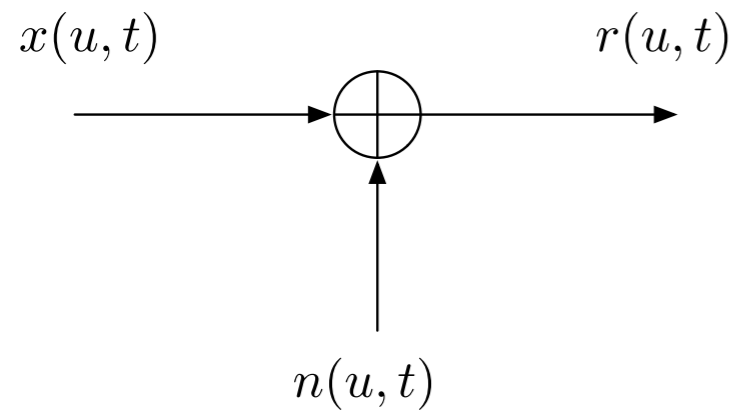


# Overview Topics

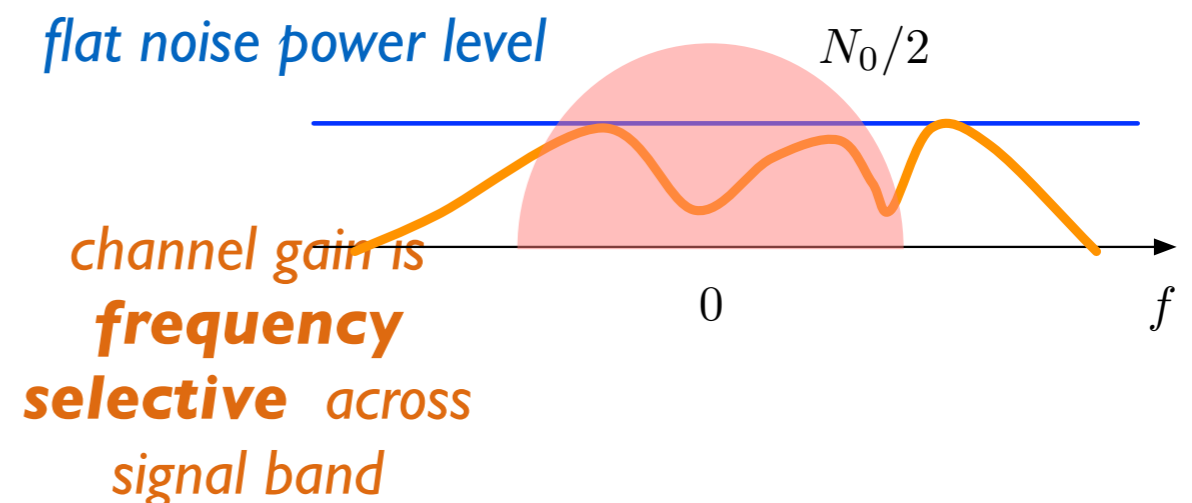
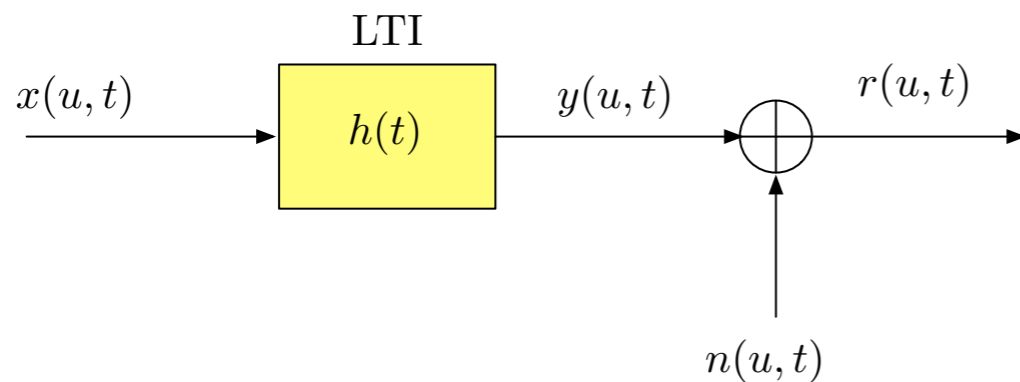
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- The digital comm system block diagram
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  - Modulations, Channels, Soft vs. Hard Decision Information
- Performance measures
- Overview of Coding
- **More Channels**

# More on Channel Models

## AWGN



## AWGN - Intersymbol Interference (ISI) Channel



# Channel Models

- We will focus primarily on the AWGN channel
- Several approaches to ISI-AWGN
  - Convert to many parallel, narrow, frequency channels with each having flat gain: **Orthogonal Frequency Division Multiplexing (OFDM)**
    - *Converts to many parallel AWGN channels*
  - Use a constrained receiver structure such as a linear filter to try to invert ISI effects: **(Linear) Channel Equalization**
  - Do optimal data detection with ISI channel modeled: **MAP Sequence/Symbol Detection — Viterbi for hard-out and Forward-Backward Algorithm for soft-out**
    - *We will learn these algorithms as part of the coding material*